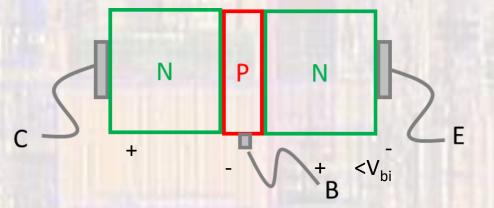
Bipolar Transistors

Last updated 1/10/24

BJT Cutoff

- Cutoff NPN
 - B-E junction < V_{bi}, C-B junction reverse biased



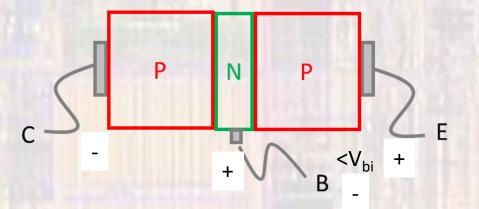
- Both diodes are 'off'
- No current flow

$$I_{C} = I_{E} = I_{S} \left[e^{\left(\frac{V_{BE}}{nV_{T}} \right)} \right] = I_{S}$$
$$I_{B} = 0$$

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BJT Cutoff

- Cutoff PNP
 - E-B junction < V_{bi}, C-B junction reverse biased



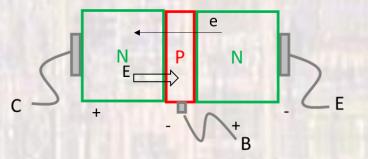
- Both diodes are 'off'
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$$I_{C} = I_{E} = I_{S} \left[e^{\left(\frac{V_{EB}}{nV_{T}} \right)} \right] = I_{S}$$
$$I_{B} = 0$$

3

BJT Forward Active

- Forward Active Mode NPN
 - B-E junction forward biased, C-B junction reverse biased
 - Collector Current
 - By convention we reference all the the currents to the collector



$$I_{C} = I_{S} \left[e^{\left(\frac{V_{BE}}{nV_{T}}\right)} \right]$$
$$I_{B} = \frac{I_{C}}{\beta} = \frac{1}{\beta} I_{S} \left[e^{\left(\frac{V_{BE}}{nV_{T}}\right)} \right]$$
$$I_{E} = \frac{1}{\alpha} I_{C} = \frac{1}{\alpha} I_{S} \left[e^{\left(\frac{V_{BE}}{nV_{T}}\right)} \right]$$

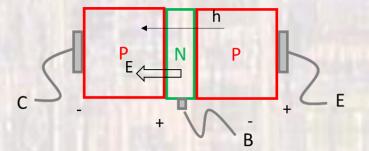
for V_{BE} > few V_T , n and I_S device dependent

$$I_C = \beta I_B$$

$$I_E = I_C + I_B$$

BJT Forward Active

- Forward Active Mode PNP
 - B-E junction forward biased, C-B junction reverse biased
 - Collector Current
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 $I_{C} = I_{S} \left[e^{\left(\frac{V_{EB}}{nV_{T}} \right)} \right]$ $I_{B} = \frac{I_{C}}{\beta} = \frac{1}{\beta} I_{S} \left[e^{\left(\frac{V_{EB}}{nV_{T}} \right)} \right]$ $I_{E} = \frac{1}{\alpha} I_{C} = \frac{1}{\alpha} I_{S} \left[e^{\left(\frac{V_{EB}}{nV_{T}} \right)} \right]$

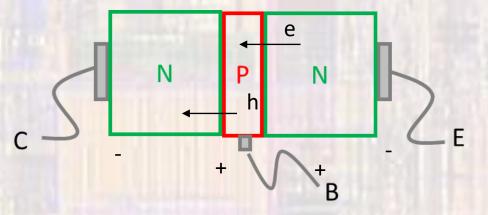
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$$I_C = \beta I_B$$

$$I_E = I_C + I_B$$

BJT Saturation

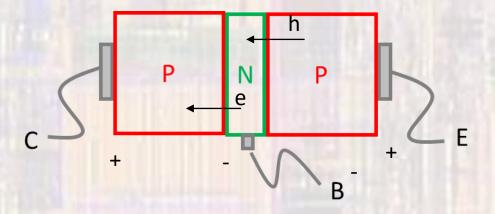
- Saturation NPN
 - B-E junction forward biased, C-B junction forward biased



- Due to the relative doping concentrations D_E > D_B > D_C, Both diodes forward biased leads to
 - Electron diffusion dominant in the B-E junction
 - Hole diffusion dominant in the B-C junction
 - but smaller than B-E electron diffusion
 - Net C-E current, linearly dependent on the C-E voltage

BJT Saturation

- Saturation PNP
 - B-E junction forward biased, C-B junction forward biased

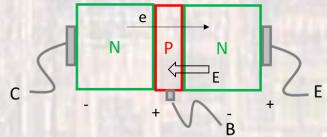


- Due to the relative doping concentrations D_E > D_B > D_C, Both diodes forward biased leads to
 - Hole diffusion dominant in the E-B junction
 - Electron diffusion dominant in the C-B junction
 - but smaller than E-B electron diffusion
 - Net E-C current, linearly dependent on the C-E voltage

BJT Reverse Active

- Reverse Active Mode NPN
 - B-E junction reverse biased, C-B junction forward biased
 - The Emitter acts as the Collector and the Collector acts as the Emitter
 - Emitter Current
 - By convention we reference all the the currents to the emitter

 $I_E = \beta I_B$



 $I_E = I_S \left[e^{\left(\frac{V_{BC}}{nV_T} \right)} \right]$

for V_{BE} > few V_T , n and I_S device dependent

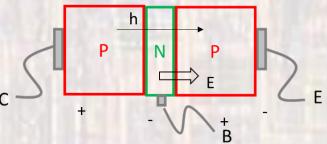
$$I_B = \frac{I_E}{\beta} = \frac{1}{\beta} I_S \left[e^{\left(\frac{V_{BC}}{nV_T} \right)} \right]$$
$$I_C = \frac{1}{\alpha} I_E = \frac{1}{\alpha} I_S \left[e^{\left(\frac{V_{BC}}{nV_T} \right)} \right]$$

 β_R is much smaller than β_F due to the relative doping of C and E

$$I_C = I_E + I_B$$

BJT Forward Active

- Forward Active Mode PNP
 - B-E junction reverse biased, C-B junction forward biased
 - The Emitter acts as the Collector and the Collector acts as the Emitter
 - Emitter Current
 - By convention we reference all the the currents to the emitter



 $I_E = I_S \left[e^{\left(\frac{V_{CB}}{nV_T} \right)} \right]$

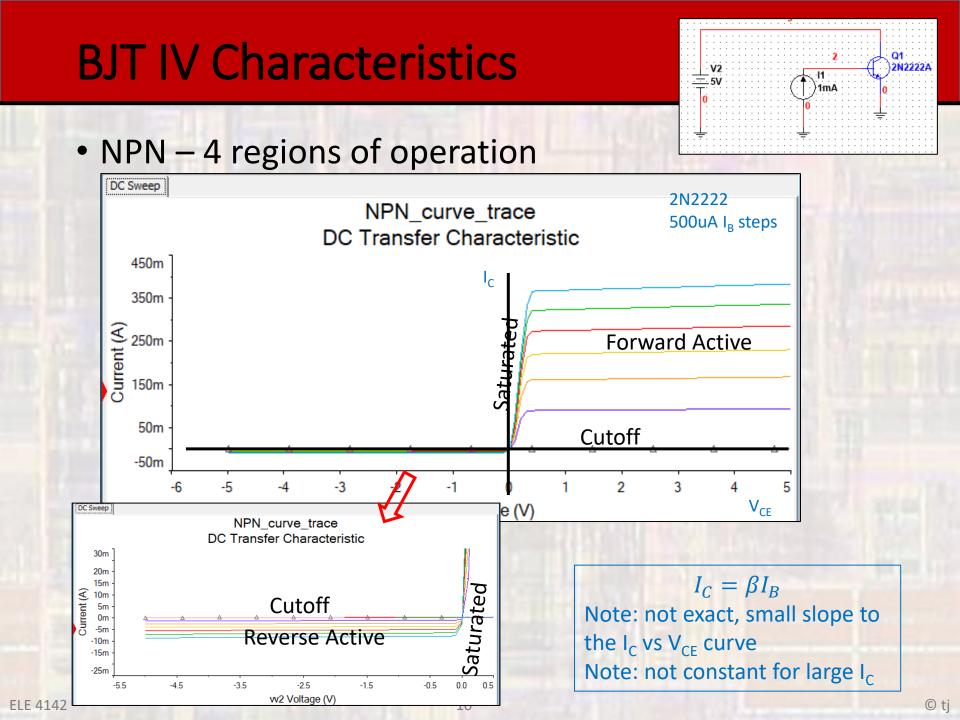
for V_{BE} > few V_T , n and I_S device dependent

$$I_B = \frac{I_E}{\beta} = \frac{1}{\beta} I_S \left[e^{\left(\frac{V_{CB}}{n_V_T} \right)} \right]$$

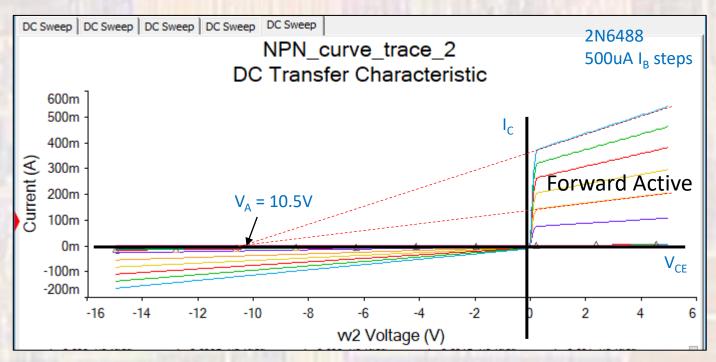
$$I_{C} = \frac{1}{\alpha} I_{E} = \frac{1}{\alpha} I_{S} \left[e^{\left(\frac{V_{CB}}{nV_{T}} \right)} \right]$$
$$I_{E} = \beta I_{B}$$

 β_R is much smaller than β_F due to the relative doping of C and E

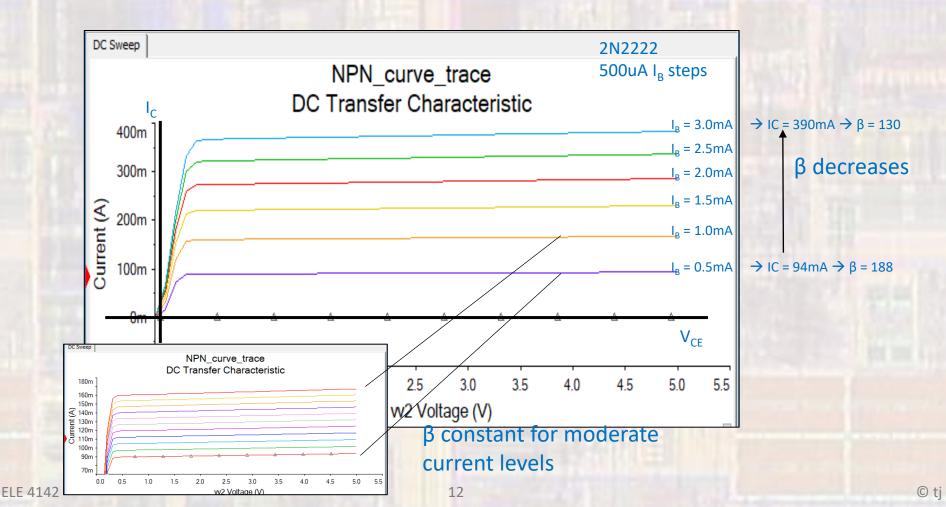
$$I_C = I_E + I_B$$



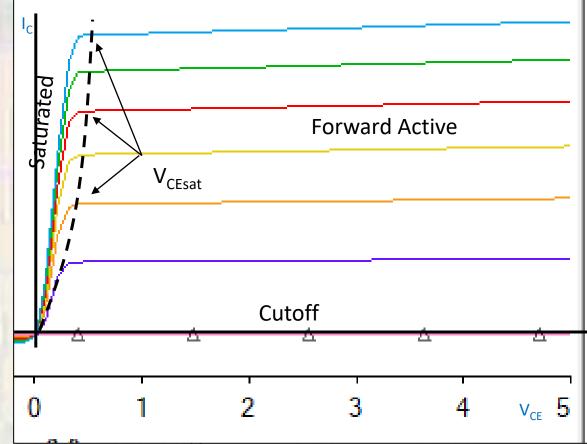
- NPN Early Voltage
 - Increasing V_{CE} → wider depletion region and greater electric field
 - \rightarrow increasing I_c
 - The curves converge on the Early Voltage, -V_A

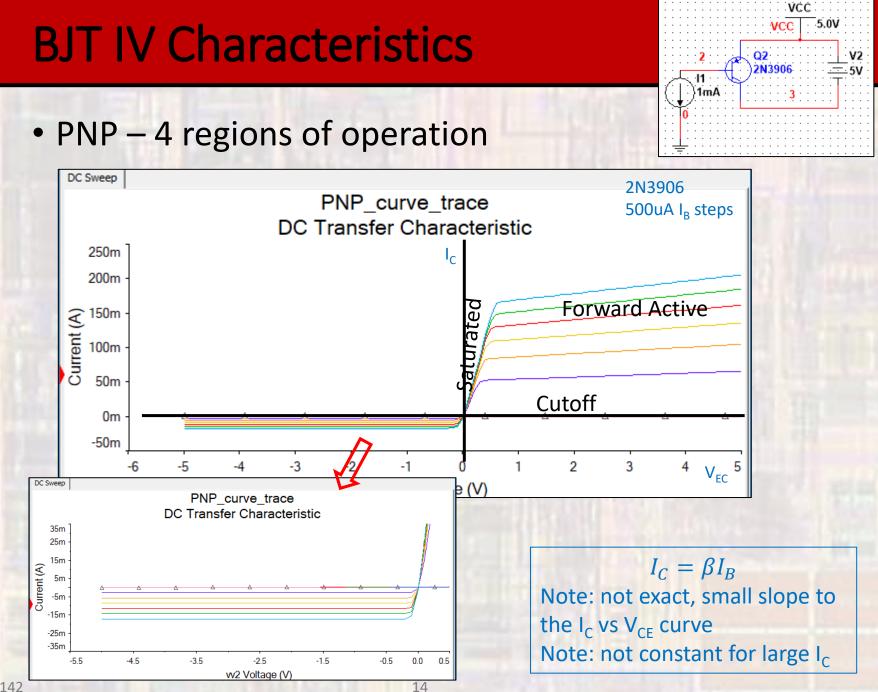


- NPN Beta variation
 - Excess carriers in the base cause \uparrow base current $\rightarrow \downarrow \beta$



- NPN V_{CEsat}
 - Saturation voltage
 - Lowest the collector will "pull down" while still providing current

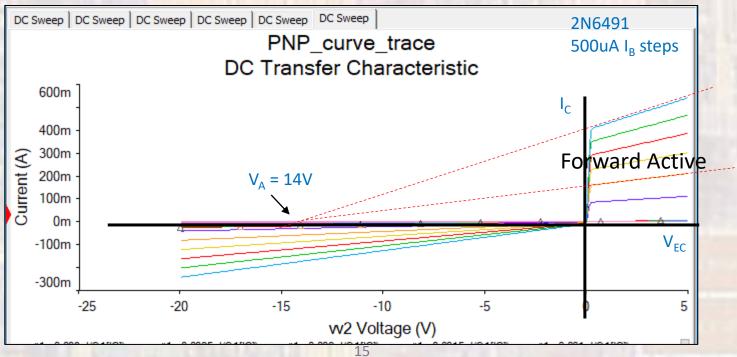




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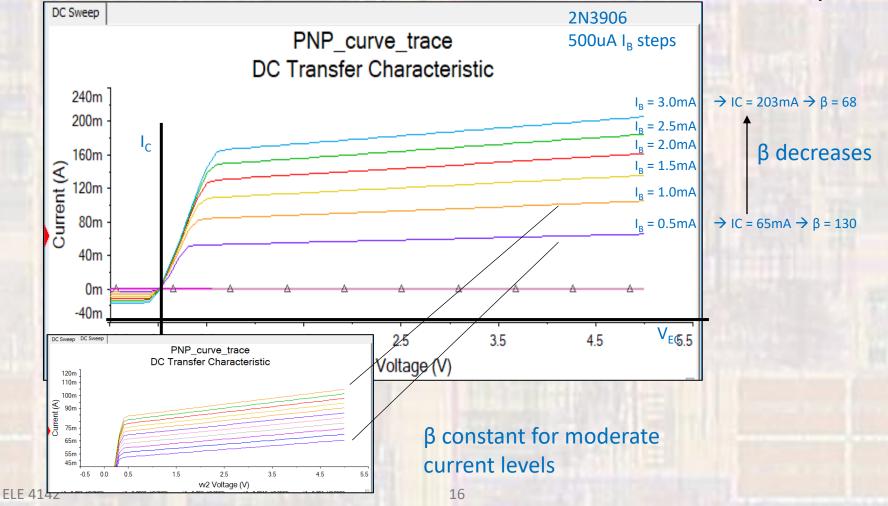
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- PNP Early Voltage
 - Increasing V_{EC} → wider depletion region and greater electric field
 - \rightarrow increasing I_c
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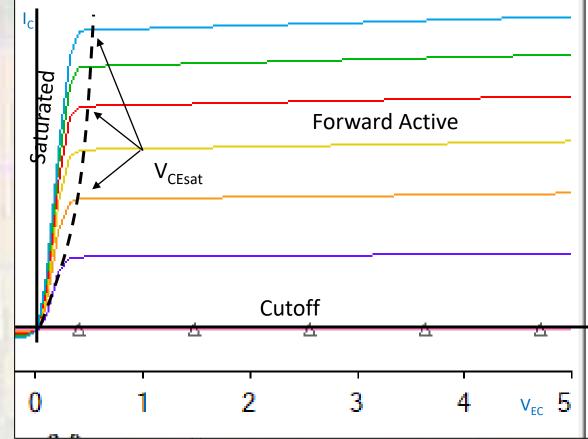
PNP – Beta variation

• Excess carriers in the base cause \uparrow base current $\rightarrow \downarrow \beta$

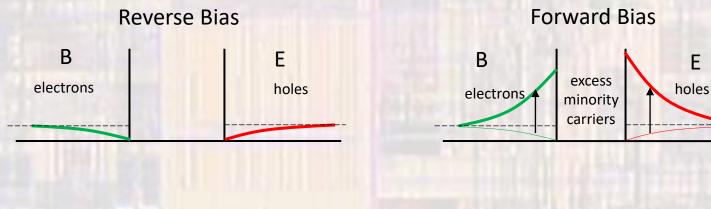


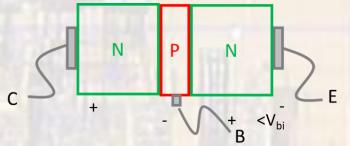
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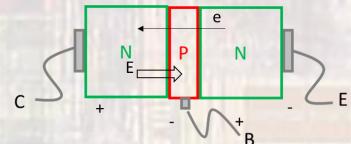
- PNP V_{CEsat}
 - Saturation voltage
 - Highest the collector will "pull up" while still providing current



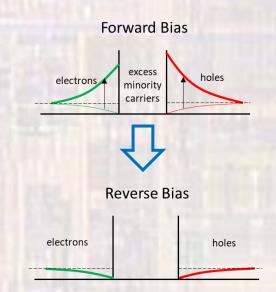
 In forward bias(B-E), carriers are traversing the depletion region and create an excess of minority carriers in the N and P regions

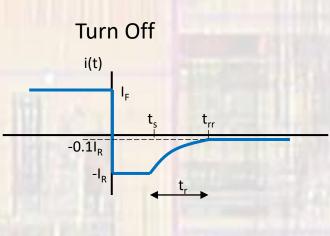




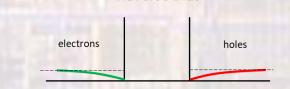


- Switching from forward bias to reverse bias
 - Excess minority carriers must be removed
 - → reverse (negative) current flow
 - Amplitude is a function of V_F and minority carrier lifetimes
 - Storage Time t_s
 - Time for concentrations to reach their OV bias level
 - Recovery Time t_r
 - Time for concentrations to reach their reverse bias level
 - Reverse Recovery Time t_{rr}
 - Sum of t_s and t_r

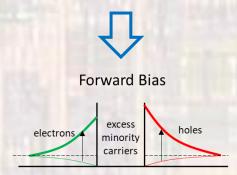


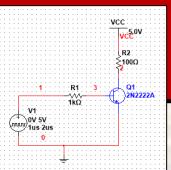


- Switching from reverse bias to forward bias
 - No excess minority carriers to be removed
 - → No storage time
 - Fast transitions

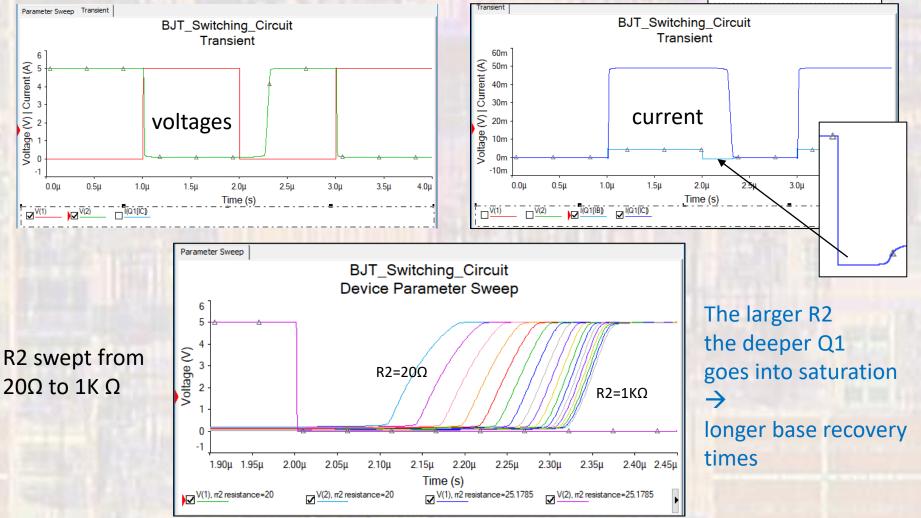


Reverse Bias





Simulation Example



Other BJTs

- Schottky Transistor
 - Combine BJT with Schottky diode
 - Used to prevent the BJT from going into saturation
 - Improved switching time
 - Can be built very efficiently on an integrated circuit

